Generalization of Wigner time delay to subunitary scattering systems

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We introduce a complex generalization of the Wigner time delay \( \tau \) for subunitary scattering systems. Theoretical expressions for complex time delays as a function of excitation energy, uniform and nonuniform loss, and coupling are given. We find very good agreement between theory and experimental data taken on microwave graphs containing an electronically variable lumped-loss element. We find that the time delay and the determinant of the scattering matrix share a common feature in that the resonant behavior in Re\[\tau\] and Im\[\tau\] serves as a reliable indicator of the condition for coherent perfect absorption (CPA). By reinforcing the concept of time delays in lossy systems this work provides a means to identify the poles and zeros of the scattering matrix from experimental data. The results also enable an approach to achieving CPA at an arbitrary frequency in complex scattering systems.

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Introduction. In this Letter we consider the general problem of scattering from a complex system by means of excitations coupled through one or more scattering channels. The scattering matrix \( S \) describes the transformation of a set of input excitations \( |\psi_{\text{in}}\rangle \) on \( M \) channels into the set of outputs \( |\psi_{\text{out}}\rangle \) as \( |\psi_{\text{out}}\rangle = S |\psi_{\text{in}}\rangle \).

A measure of how long the excitation resides in the interaction region is provided by the time delay, related to the energy derivative of the scattering phase(s) of the system. This quantity and its variation with energy and other parameters can provide useful insights into the properties of the scattering region and has attracted research attention since the seminal works by Wigner [1] and Smith [2]. A review of the theoretical aspects of time delays with an emphasis on solid state applications can be found in Ref. [3]. Various aspects of time delays have recently been shown to be of direct experimental relevance for manipulating wave fronts in complex media [4–6]. Time delays are also long known to be directly related to the density of states of open scattering systems (see discussions in Ref. [3] and more recently in Refs. [7,8]).

For the case of flux-conserving scattering in systems with no losses, the \( S \) matrix is unitary and its eigenvalues are phases \( e^{i\theta_a}, a = 1, 2, \ldots, M \). These phases are functions of the excitation energy \( E \) and one can then define several different measures of time delay (see, e.g., Refs. [3,9]), such as partial time delays associated with each channel \( \tau_a = \frac{d\theta_a}{dE} \), the proper time delays which are the eigenvalues of the Wigner-Smith matrix \( \hat{Q} = i\hbar S' S \), and the Wigner delay time which is the average of all the partial time delays \( \tau_W = \frac{1}{M} \sum_{a=1}^{M} \tau_a = \frac{1}{2\hbar} \text{Tr}(\hat{Q}) \).

A rich class of systems in which the properties of various time delays enjoyed thorough theoretical attention is scattering of short-wavelength waves from classically chaotic systems, e.g., billiards with ray-chaotic dynamics or particles on graphs, e.g., such as considered in Ref. [10]. Various examples of chaotic wave scattering (quantum or classical) have been observed in nuclei, atoms, molecules, ballistic two-dimensional electron gas billiards, and most extensively in microwave experiments [11–16]. In such systems time delays have been measured starting from the pioneering work [17], followed over the last three decades by measurement of the statistical properties of time delays through random media [18,19] and microwave billiards [20]. A Wigner time delay for an isolated resonance described by an \( S \)-matrix pole at a complex energy \( E_0 - i\Gamma \) has a value of \( Q = 2\hbar/\Gamma \) on resonance, hence studies of the imaginary part of the \( S \)-matrix poles probe one aspect of time delays [21–26]. In the meantime, the Wigner-Smith operator (WSO) was utilized to identify minimally dispersive principal modes in coupled multimode systems [27,28]. A similar idea was used to create particlelike scattering states as eigenstates of the WSO [4,29,30]. A generalization of the WSO allowed maximal focus on, or maximal avoidance of, a specific target inside a multiple scattering medium [6,31].

Time delays in wave-chaotic scattering are expected to be extremely sensitive to variations of the excitation energy and scattering system parameters, and will display universal fluctuations when considering an ensemble of scattering systems with the same general symmetry. The universality of fluctuations allows them to be efficiently described using the theory of random matrices [9,32–40]. Alternative theoretical
treatments of time delays in chaotic scattering systems successfully adopted a semiclassical approach (see Ref. [7] and references therein).

Despite the fact that the standard theory of wave-chaotic scattering deals with perfectly flux-preserving systems, in any actual realization such systems are inevitably imperfect, hence absorbing, and theory needs to take this aspect into account [41]. Interestingly, studying scattering characteristics in a system with weak uniform (i.e., spatially homogeneous) losses may even provide a possibility to extract time delays characterizing an idealized system without losses. This idea has been experimentally realized already in Ref. [17] which treated the effect of subunitary scattering by means of the unitary deficit of the $S$ matrix. In this case consider the $Q$ matrix defined through the relation $S' S = 1 - (\gamma \Delta / 2\pi)Q_{UL}$, where $\gamma$ is the dimensionless “absorption rate” and $\Delta$ is the mean spacing between modes of the closed system. In the limit of vanishing absorption rate $\gamma \to 0$ such $Q_{UL}$ can be shown to coincide with the Wigner-Smith time-delay matrix for a lossless system, but formally one can extend this as a definition of $Q$ for any $\gamma > 0$. Note that this version of a time delay is always real and positive. Various statistical aspects of time delays in such and related settings were addressed theoretically in Refs. [42–45].

Experimental data are often taken on subunitary scattering systems and a straightforward use of the Wigner time-delay definition yields a complex quantity. In addition, both the real and imaginary parts acquire both negative and positive values, and they show a systematic evolution with energy or other parameters of the scattering system. This clearly calls for a detailed theoretical understanding of this complex generalization of the Wigner time delay. It is necessary to stress that many possible definitions of time delays which are equivalent or directly related to each other in the case of a lossless flux-conserving systems can significantly differ in the presence of flux losses, either uniform or spatially localized. In the present Letter we focus on a definition that can be directly linked to the fundamental characteristics of the scattering matrix—its poles and zeros in the complex energy plane—making it useful for fully characterizing an arbitrary scattering system. Note that $S$-matrix poles have been objects of long-standing theoretical [46–54] and experimental [21–23,25] interest in chaotic wave scattering, whereas $S$-matrix zeros started to attract research attention only recently [26,55–63].

**Complex Wigner time delay.** In our exposition we use the framework of the so-called “Heidelberg approach” to wave-chaotic scattering reviewed from different perspectives in Refs. [64–66]. Let $H$ be the $N \times N$ Hamiltonian which is used to model the closed system with ray-chaotic dynamics, $W$ denoting the $N \times M$ matrix of coupling elements between the $N$ modes of $H$ and the $M$ scattering channels, and by $A$ the $N \times L$ matrix of coupling elements between the modes of $H$ and the $L$ localized absorbers, modeled as $L$ absorbing channels [67]. The total unitary $S$ matrix, of size $(M + L) \times (M + L)$ describing both the scattering and absorption on equal footing, has the following block form (see, e.g., Refs. [56]),

$$S(E) = \begin{pmatrix} 1_M - 2\pi i W^\dagger D^{-1}(E)W & -2\pi i W^\dagger D^{-1}(E)A \\ -2\pi i A^\dagger D^{-1}(E)W & 1_L - 2\pi i A^\dagger D^{-1}(E)A \end{pmatrix},$$  

(1)

where we defined $D(E) = E - H + i(\Gamma_W + \Gamma_A)$ with $\Gamma_W = \pi W W^\dagger$ and $\Gamma_A = \pi A A^\dagger$.

The upper left diagonal $M \times M$ block of $S(E)$ is the experimentally accessible subunitary scattering matrix and is denoted as $S_\gamma(E)$. The presence of uniform-in-space absorption with strength $\gamma$ can be taken into account by evaluating the $S$-matrix entries at a complex energy: $S(E + i\gamma) := S_\gamma(E)$. The determinant of such a subunitary scattering matrix $S_\gamma(E)$ is then given by

$$\det S_\gamma(E) = \det S(E + i\gamma)$$  

(2)

$$= \frac{\det [E - H + i(\gamma + \Gamma_W - \Gamma_A)]}{\det [E - H + i(\gamma + \Gamma_A + \Gamma_W)]}$$  

(3)

$$= \prod_{n=1}^{N} \frac{E + i\gamma - z_n}{E + i\gamma - \bar{z}_n},$$  

(4)

In the above expression we have used that the $S$-matrix zeros $z_n$ are complex eigenvalues of the non-self-adjoint (non-Hermitian) matrix $H + i(\Gamma_W - \Gamma_A)$, whereas the poles $\bar{z}_n = E_n - i\Gamma_n$ with $\Gamma_n > 0$ are complex eigenvalues of yet another non-Hermitian matrix $H - i(\Gamma_W + \Gamma_A)$, frequently called in the literature “the effective non-Hermitian Hamiltonian” [9,46,54,65,66,68]. Note that when localized absorption is absent, i.e., $\Gamma_A = 0$, the zeros $z_n$ and poles $\bar{z}_n$ are complex conjugates of each other, as a consequence of $S$-matrix unitarity for real $E$ and no uniform absorption $\gamma = 0$. Extending to locally absorbing systems the standard definition of the Wigner delay time as the energy derivative of the total phase shift we now deal with a complex quantity:

$$\tau(E; A, \gamma) := -\frac{i}{M} \frac{\partial}{\partial E} \log \det S_\gamma(E)$$  

(5)

$$= \text{Re} \tau(E; A, \gamma) + i \text{Im} \tau(E; A, \gamma),$$  

(6)

$$\text{Re} \tau(E; A, \gamma) = \frac{1}{M} \sum_{n=1}^{N} \left[ \frac{\text{Im} z_n - \gamma}{(E - \text{Re} z_n)^2 + (\text{Im} z_n - \gamma)^2} + \frac{\Gamma_n + \gamma}{(E - E_n)^2 + (\Gamma_n + \gamma)^2} \right],$$  

(7)

$$\text{Im} \tau(E; A, \gamma) = -\frac{1}{M} \sum_{n=1}^{N} \left[ \frac{E - \text{Re} z_n}{(E - \text{Re} z_n)^2 + (\text{Im} z_n - \gamma)^2} - \frac{E - E_n}{(E - E_n)^2 + (\Gamma_n + \gamma)^2} \right].$$  

(8)
Equation (7) for the real part is formed by two Lorentzians for each mode of the closed system, potentially with different signs. This is a striking difference from the case of the flux-preserving system in which the conventional Wigner time delay is expressed as a single Lorentzian for each resonance mode [69]. Namely, the first Lorentzian is associated with the \( n \)th zero while the second is associated with the corresponding pole of the scattering matrix. The widths of the two Lorentzians are controlled by system scattering properties, and when \( \text{Im} z_n \to \gamma \pm 0 \) the first Lorentzian in Eq. (7) acquires the divergent, delta-functional peak shape, of either positive or negative sign, centered at \( E = \text{Re} z_n \). Note that the first term in Eq. (8) changes its sign at the same energy value. These properties are indicative of the "perfect resonance" condition, with divergence in the real part of the Wigner time delay signaling the wave or (quantum) particle being perpetually trapped in the scattering environment. In different words, the energy of the incident wave or particle is perfectly absorbed by the system due to the finite losses.

The pair of equations (7) and (8) forms the main basis for our consideration. In particular, we demonstrate in Supplemental Material Sec. I [70] that in the regime of well-resolved resonances Eqs. (7) and (8) can be used for extracting the positions of both poles and zeros in the complex plane from experimental measurements, provided the rate of uniform absorption \( \gamma \) is independently known. We would like to stress that in general the two Lorentzians in (7) are centered at different energies because generically the pole position \( E_n \) does not coincide with the real part of the complex zero \( \text{Re} z_n \).

From a different angle it is worth noting that there is a close relation between the objects of our study and the phenomenon of the so-called coherent perfect absorption (CPA) which attracted considerable attention in recent years, both theoretically and experimentally [60,62,71–73]. Namely, the above-discussed match between the uniform absorption strength and the imaginary part of scattering matrix zero \( \gamma = \text{Im} z_n \) simultaneously ensures the determinant of the scattering matrix to vanish [see Eq. (4)]. This is only possible when \( |\psi_{\text{out}}\rangle = 0 \) despite the fact that \( |\psi_{\text{in}}\rangle \neq 0 \), which is a manifestation of CPA (see, e.g., Refs. [55,56]).

**Experiment.** We focus on experiments involving microwave graphs [13,62,74,75] for a number of reasons. First, they provide for complex scattering scenarios with well-isolated modes amenable to a detailed analysis. We thus avoid
the complications of interacting poles and related interference effects [76]. Graphs also allow for convenient parametric control such as variable lumped lossy elements, variable global loss, and breaking of time-reversal invariance. We utilize an irregular tetrahedral microwave graph formed by coaxial cables and T-junctions, having $M = 2$ single-mode ports, and broken time-reversal invariance. A voltage-controlled variable attenuator is attached to one internal node of the graph [see Fig. 1(a)], providing for a variable lumped loss $(L = 1$, the control variable $\Gamma_A$). The nodes involving connections of the graph to the network analyzer, and the graph to the lumped loss, are made up of a pair of T-junctions. The coaxial cables and T-junctions have a roughly uniform and constant attenuation produced by dielectric loss and conductor loss, which is parametrized by the uniform loss parameter $\gamma$. The two-port graph has a total electrical length of $L_e = 3.89$ m, a mean mode spacing of $\Delta = c/2L_e = 38.5$ MHz, and a Heisenberg time $\tau_0 = 2\pi/\Delta = 163$ ns. The graph has equal coupling on both ports, characterized by a nominal value of $T_A = 0.9450$ at a frequency of 2.6556 GHz [77].

Comparison of theory and experiments. Figure 1 shows the evolution of a complex time delay for a single isolated mode of the $M = 2$ port tetrahedral microwave graph as $\Gamma_A$ is varied. The complex time delay is evaluated as in Eq. (5) based on the experimental $S(f)$ data, where $f$ is the microwave frequency, a surrogate for energy $E$. Note that the (calibrated) measured $S$-parameter data are directly used for calculation of the complex time delay without any data preprocessing. The resulting real and imaginary parts of the time delay vary systematically with frequency, adopting both positive and negative values, depending on frequency and lumped loss in the graph. The full evolution animated over varying lumped loss is available in the Supplemental Material [70]. These variations are well described by the theory given above.

Figures 1(d) and 1(e) clearly demonstrate that two Lorentzians are required to correctly describe the frequency dependence of the real part of the time delay. The two Lorentzians have different widths in general, given by the values of $\text{Im} \tau_n - \gamma$ and $\Gamma_n + \gamma$, and in this case the Lorentzians also have opposite signs. The frequency dependence of the imaginary part of the time delay also requires two terms, with the same parameters as for the real part, to be correctly described. The data in Fig. 1(b) also reveal that Re$[\tau]$ goes to very large positive values and suddenly changes sign to large negative values at a critical amount of local loss. For another attenuation setting of the same mode it was found that the maximum delay time was 337 times the Heisenberg time, showing that the signal resides in the scattering system for a substantial time.

The measured complex time delay as a function of frequency can be fit to Eqs. (7) and (8) to extract the corresponding pole and zero location for the $S$ matrix. The method to perform this fit is described in Supplemental Material Sec. I [70]. The fitting parameters are $\text{Re} \tau_n$ and $\text{Im} \tau_n - \gamma$ for the zero, and $\Gamma_n + \gamma$ for the pole. Note that the Re$[\tau(f)]$ and Im$[\tau(f)]$ data are fit simultaneously, and constant offsets $C_R$ and $C_i$ are added to each fit.

Figure 2 summarizes the parameters required to fit the experimental complex time delay versus frequency (shown in Fig. 1) as the localized loss due to the variable attenuator in the graph is increased. The significant feature here is the zero crossing of $\text{Im} \tau_n - \gamma$ at frequency $f = f_{\text{CPA}}$, which reaches its minimum at the zero-crossing point. The inset shows the evolution of $\text{Re} \tau_n$ and $E_n = \text{Re} \epsilon_n$ with attenuation. (b) Evolution of complex zero and pole of a single mode of the graph in the complex frequency plane as a function of $\Gamma_A$. The black crosses are the initial state of the zero and pole at the minimum attenuation setting. The insets show the details of the complex zero and pole migrations.

![Figure 2](image-url)

FIG. 2. (a) Fitted parameters $\text{Im} \tau_n - \gamma$ and $\Gamma_n + \gamma$ for the complex Wigner time delay from graph experimental data. Also shown is the evolution of $|\det(S)|$ at the specific frequency of interest, $f_{\text{CPA}}$, which reaches its minimum at the zero-crossing point. The inset shows the evolution of $\text{Re} \tau_n$ and $E_n$ with attenuation. (b) Evolution of complex zero and pole of a single mode of the graph in the complex frequency plane as a function of $\Gamma_A$. The black crosses are the initial state of the zero and pole at the minimum attenuation setting. The insets show the details of the complex zero and pole migrations.
evaluate the uniform loss strength $\gamma$ to be $3.73 \times 10^{-3} \text{GHz}$ (see Supplemental Material Sec. III [70]).

Figure 2(b) summarizes the locations of the $S$-matrix pole $\varepsilon_n$ and zero $z_n$ of the single isolated mode of the microwave graph in the complex frequency plane as the localized loss is varied. When the $S$ matrix zero crosses the $\text{Im} z_n = \gamma$ value, one has the traditional signature of CPA. Note from Fig. 2 that the real parts of the zero and pole do not coincide and in fact move away from each other as localized loss is increased.

Discussion. It should be noted that the occurrence of a negative real part of the time delay is an inevitable consequence of subunitary scattering, and is also expected for particles interacting with attractive potentials [78].

The imaginary part of the time delay was in the past discussed in relation to changes in the scattering unitary deficit with frequency [30]. Another approach to defining a complex time delay has been recently suggested to be based on essentially calculating the time delay of the signal which comes out of the system without being absorbed [73]. It should be noted that this ad hoc definition of time delay is not simply related to the poles and zeros of the $S$ matrix. Moreover, a closer inspection shows that such a definition of a complex time delay tacitly assumes that the real parts of the pole and zero are identical. According to our theory such an assumption is incompatible with a proper treatment of localized loss.

We emphasize that the correct knowledge of the locations of the poles and zeros is essential for reconstructing the scattering matrix over the entire complex energy plane through Weierstrass factorization [79]. Through graph simulations presented in Supplemental Material Sec. VII [70] we demonstrate that the complex time-delay theory presented here also works for time-reversal invariant systems, and for systems with variable uniform absorption strength $\gamma$. Our results therefore establish a systematic procedure to find the $S$-matrix zeros and poles of isolated modes of a complex scattering system with an arbitrary number of coupling channels, symmetry class, and arbitrary degrees of both global and localized loss.

Recent work has demonstrated CPA in disordered and complex scattering systems [60,62]. It has been discovered that one can systematically perturb such systems to induce CPA at an arbitrary frequency [73,80], and this enables a remarkably sensitive detector paradigm [73]. These ideas can also be applied to optical scattering systems where measurements of the transmission matrix are possible [81]. Here, we have uncovered a general formalism in which to understand how CPA can be created in an arbitrary scattering system. In particular, this work shows that both the global loss ($\gamma$), localized loss centers, or changes to the spectrum can be independently tuned to achieve the CPA condition.

Future work includes treating the case of overlapping modes, and the development of theoretical predictions for the statistical properties of both the real and imaginary parts of the complex time delay in chaotic and multiple scattering subunitary systems.

Conclusions. We have introduced a complex generalization of the Wigner time delay which holds both for arbitrary uniform (global) and spatially-localized loss, and directly relates to poles and zeros of the scattering matrix in the complex frequency (or energy) plane. Based on that we developed theoretical expressions for a complex time delay as a function of frequency (or energy), and found very good agreement with experimental data on a subunitary complex scattering system. The time delay and $\text{det}(S)$ share a common feature in that CPA and the divergence of $\text{Re}[\tau]$ and $\text{Im}[\tau]$ coincide. This work opens a window on time delays in lossy systems, enabling the extraction of complex zeros and poles of the $S$-matrix from data.

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This way of modeling the localized absorbers as additional 


[67] This way of modeling the localized absorbers as additional scattering channels is close in spirit to the so-called dephasing lead model of decoherence introduced in M. Böttiker, Role of quantum coherence in series resistors, Phys. Rev. B 33, 3020 (1986); and further developed in P. W. Brouwer and C. W. J. Beenakker, Voltage-probe and imaginary-potential models for dephasing in a chaotic quantum dot, ibid. 55, 4695 (1997).


[70] See Supplemental Material at http://link.aps.org/supplemental/10.1103/PhysRevE.103.L050203 for the details of extracting poles and zeros from data, the sign convention used for the scattering matrix frequency evolution, evaluation of the system uniform loss strength γ, discussions about effects of neighboring resonances on fitting to the complex time delay, further details about CPA and complex time delay, connections to earlier work on negative real time delay and imaginary time delay, simulations of time-reversal invariant graphs and evaluation of complex time delay with varying uniform loss, and for animations of time delay evolution with a variation of lumped loss [experiment Figs. 1(b) and 1(c)] or uniform loss [simulation Figs. S4(b) and S4(c)].


[77] The coupling strength Tc is determined by the value of the radiation S matrix (Tc = 1 − |S|2). The radiation S is measured when the graph is replaced by 50Ω loads connected to the three output connectors of each node attached to the network analyzer test cables.


